



Dual H-Bridge Stepper Motor Driver

1 Features

- Dual H-Bridge Current Control Motor Driver
 - One Stepper Motor
 - Low MOSFET ON-Resistance:
 - HS + LS 1.0Ω(Typical, 25°C)
- Output Current Capability (at V_M = 5V, 25°C)
 eTSSOP Package:
 - 1.0A RMS, 1.2A Peak per H-Bridge
 - QFN Package:
 - 0.9A RMS, 1.2A Peak per H-Bridge
- Wide Power Supply Voltage: 2.5V to 10.8V
- Integrated Current Regulation
- Easy Pulse-Width-Modulation (PWM) Interface
- Protection Features
 - VM Undervoltage Lockout (UVLO)
 - Overcurrent Protection (OCP)
 - Thermal Shutdown (TSD)
 - Fault Indication Pin(nFAULT)

2 Applications

- Point-of-Sale Printers
- Video Security Cameras
- Office Automation Machines
- Gaming Machines
- Robotics
- Battery-Powered Toys

3 Description

The GD30DR3820 provides a dual-bridge steppwe motor driver solution for cameras, printers, toys, robotics and other mechatronic applications.

The device has two H-bridge drivers, drive a bipolar stepper motor. The output driver block consists of Nchannel power MOSFETs configured as an H-bridge to drive the motor winding. An internal charge pump generates gate drive voltages.

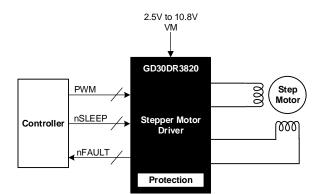
The GD30DR3820 device has two PWM(IN/IN) input interface. Internal shutdown functions with a fault output pin are provided for overcurrent protection, short circuit protection, undervoltage lockout, and overtemperature. In addition, GD30DR3820 also supports low-power sleep mode.

The GD30DR3820 is packaged in a 16-pin eTSSOP and 16-pin QFN package.

Device Information¹

PART NUMBER	PACKAGE	BODY SIZE (NOM)
GD30DR3820	eTSSOP (16)	5.00 mm × 6.40 mm
GD30DK3620	QFN (16)	3.00 mm × 3.00 mm

1. For packaging details, see *Package Information* section.



Simplified Application Schematic



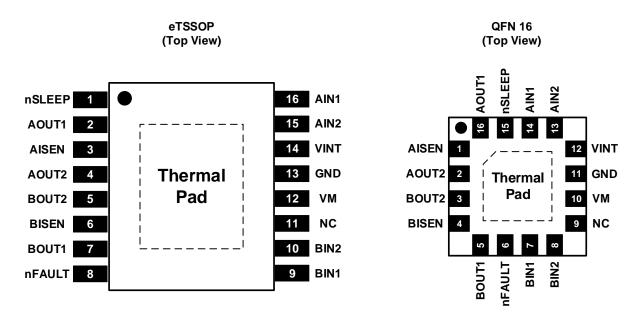
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4 Device Overview

4.1 Pinout and Pin Assignment



4.2 Pin Description

F	PIN NUMBE	R		FUNCTION
NAME	eTSSOP	QFN		FUNCTION
nSLEEP	1	15	1	Sleep mode input. Logic high to enable device; Logic low to enter low-
		10		power sleep mode and reset all internal logic; Internal pulldown.
AOUT1	2	16	0	Bridge A output 1. Connect to motor winding A.
AISEN	3	1	I/O	Bridge A ground or Isense. Connect to current sense resistor for bridge
AIGEN	5	Ι	1/0	A or GND if current control not needed.
AOUT2	4	2	0	Bridge A output 2. Connect to motor winding A.
BOUT2	5	3	0	Bridge B output 2. Connect to motor winding B.
BISEN	6	4	I/O	Bridge B ground or Isense. Connect to current sense resistor for bridge
DISEN	0	4	1/0	B or GND if current control not needed.
BOUT1	7	5	0	Bridge B output 1. Connect to motor winding B.
nFAULT	8	6	OD	Fault output. Logic low when in fault condition (overtemperature,
IIFAULI	0	0	00	overcurrent, undervoltage lockout).
BIN1	9	7	I	Bridge B input 1. Controls the state of BOUT1. Internal pulldown.
BIN2	10	8	I	Bridge B input 2. Controls the state of BOUT2. Internal pulldown.
NC	11	9		No connection pin.
VM	12	10	Р	Power supply. Connect to motor supply. A 10uF(minimum) ceramic
VIVI	12	10		bypass capacitor to GND is recommended.
GND	13	11	Р	Device ground. This pin must be connected to the PCB ground.



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		10	5	
VINT	14	12	P	Internal regulator. Bypass to GND with a 2.2uF, 6.3V capacitor.
AIN2	15	13	I	Bridge A input 2. Controls the state of AOUT2. Internal pulldown.
AIN1	16	14	Ι	Bridge A input 1. Controls the state of AOUT1. Internal pulldown.
GND	Thormal	Thermal Thermal P		Power ground, connect to board ground, use large ground plane for good
GND	GND Thermal			thermal dissipation, and multiple nearby vias connecting those planes.

1. I = input, O = output, I/O= input/output, OD =open drain, P = power.



5 Parameter Information

5.1 Absolute Maximum Ratings

Exceeding the operating temperature range (unless otherwise noted)^{1,2}

SYMBOL	PARAMETER	MIN	MAX	UNIT
VM	Power supply voltage	-0.3	12	V
VINT	Internal regulator	-0.3	3.8	V
	Logic input pin voltage	-0.5	7.0	V
	Peak motor drive output current	Internal	ly limited	Α
TJ	Operating junction temperature	-40	150	°C
T _{stg}	Storage temperature	-60	150	°C

1. Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

2. All voltage values are with respect to network ground terminal.

5.2 Recommended Operation Conditions

SYMBOL ^{1,2}	PARAMETER			ТҮР	MAX	UNIT
VM	Motor power supply voltage ³	Motor power supply voltage ³			10.8	V
VLOGIC	Logic level input voltage		0		5.5	V
	Motor RMS current⁴	eTSSOP16 Package	0		1.0	А
IRMS	Motor RMS current	QFN16 Package	0		0.9	А
f _{pwm}	Externally applied PWM frequency		0		200	kHz
T _A	Operating ambient temperature	9 ⁴	-40		85	°C

1. The device is not guaranteed to function outside of its operating conditions.

2. Refer to the Application Information section for further information.

3. Note that when VM is below 5 V, $R_{DS(ON)}$ increases and maximum output current is reduced.

4. Power dissipation and thermal limits must be observed.

5.3 Electrical Sensitivity

SYMBOL	CONDITIONS	VALUE	UNIT
V _{ESD(HBM)}	Human-body model (HBM), ANSI/ESDA/JEDEC JS-001-2017 ¹	±3000	V
V _{ESD(CDM)}	Charge-device model (CDM), ANSI/ESDA/JEDEC JS-002-2022 ²	±1500	V

1. JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

2. JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



5.4 Electrical Characteristics

$T_A = 25^{\circ}C$, unless otherwise noted.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
POWER	SUPPLY (VM, VINT)					
VM	VM operating voltage		2.5		10.8	V
I _{∨M}	VM operating supply current	V _M =5V, xINx low, nSLEEP high		0.6	1.0	mA
IVMQ	VM sleep mode supply current	V _M =5V, nSLEEP low		30	95	nA
t _{sleep}	Sleep time	nSLEEP low to sleep mode		10		μs
t _{wake}	Wake-up time	nSLEEP high to output transition		155		μs
t _{on}	Turn-on time	V_M > V_{UVLO} to output transition		25		μs
VINT	Internal regulator voltage	V _M = 5V	3.2	3.5	3.8	V
M	VM undervoltage lockout	V _M rising			2.5	V
Vuvlo	voltage	V _M falling			2.4	V
LOGIC-LI	EVEL INPUTS (AIN1, AIN2, B	IN1, BIN2, nSLEEP)				
		xINx	0		0.7	V
VIL	Input low voltage	nSLEEP	0		0.5	V
		xINx	1.8		5.5	V
Ven_fall	Input high voltage	nSLEEP	2.3		5.5	V
V _{HYS}	Input logic hysteresis		200	300	500	mV
lı∟	Input low current	VIN = 0	-5		5	μA
Iн	Input high current	VIN = 5V			50	μA
RPD	Pulldown resistance			100		KΩ
t _{DEG}	Input deglitch time				600	ns
T _{PROP}	Propagation delay INx to OUTx	V _M = 5V			1.2	μs
CONTRO	L OUTPUT (nFAULT)					
Vol	Output logic low voltage	I ₀ = 5mA			0.1	V
Іон	Output logic high leakage	V ₀ = 3.3V	-1		1	μA
MOTOR [DRIVER OUTPUTS(AOUT1, A	OUT2, BOUT1, BOUT2)				
		$V_M = 5V$, $I_O = 0.2A$, $T_J = -40^{\circ}C^1$		400		mΩ
		V _M = 5V, I _O = 0.2A, T _J = 20°C		550		mΩ
	High side EET on registeres	$V_{M} = 5V, I_{O} = 0.2A, T_{J} = 85^{\circ}C^{1}$		650		mΩ
Rds(on)	High-side FET on resistance	V_{M} = 2.7V, I_{O} = 0.2A, T_{J} = -40°C ¹		650		mΩ
		V _M = 2.7V, I _O = 0.2A, T _J =25°C		800		mΩ
		V _M = 2.7V, I _O = 0.2A, T _J = 85°C ¹		900		mΩ
		$V_{M} = 5V, I_{O} = 0.2A, T_{J} = -40^{\circ}C^{1}$		350		mΩ
	Low side FET on registeres	V _M = 5V, I _O = 0.2A, T _J = 20°C		450		mΩ
	Low-side FET on resistance	$V_{M} = 5V, I_{O} = 0.2A, T_{J} = 85^{\circ}C^{1}$		550		mΩ
		$V_{M} = 2.7V, I_{O} = 0.2A, T_{J} = -40^{\circ}C^{1}$		650		mΩ



Electrical Characteristics(continued)

$T_A = 25^{\circ}C$ (unless otherwise noted).

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
D		V _M = 2.7V, I _O = 0.2A, T _J =25°C		800		mΩ
Rds(on)	Low-side FET on resistance	V_{M} = 2.7V, I_{O} = 0.2A, T_{J} = 85°C ¹		900		mΩ
IOFF	OFF-state leakage current	V _M = 5V	-1		1	μA
t _{RISE}	Output rise time	V _M = 5V, 16Ω to GND, 10% to 90%VM		70		ns
t _{FALL}	Output fall time	V _M = 5V, 16Ω to GND, 10% to 90%VM		80		ns
TDEAD	Dead time ²	V _M = 5V		450		ns
PWM CU	RRENT CONTROL (AISEN, B	ISEN)				
V _{TRIP}	xISEN trip voltage		160	200	240	mV
toff	Current control constant off time	Internal PWM constant off time		23		μs
PROTECT	TION CIRCUITS					
IOCP	Overcurrent protection trip level		1.2			А
t _{DEG}	Overcurrent de-glitch time			1		μs
tocr	Overcurrent protection retry time			1.4		ms
T_{TSD}^{1}	Thermal shutdown temperature		150			°C
T _{HYS}	Thermal shutdown hysteresis			20		°C

1. Not tested in production; based on design and characterization data.

2. Internal dead time. External implementation is not necessary.



6 Functional Description

6.1 Block Diagram

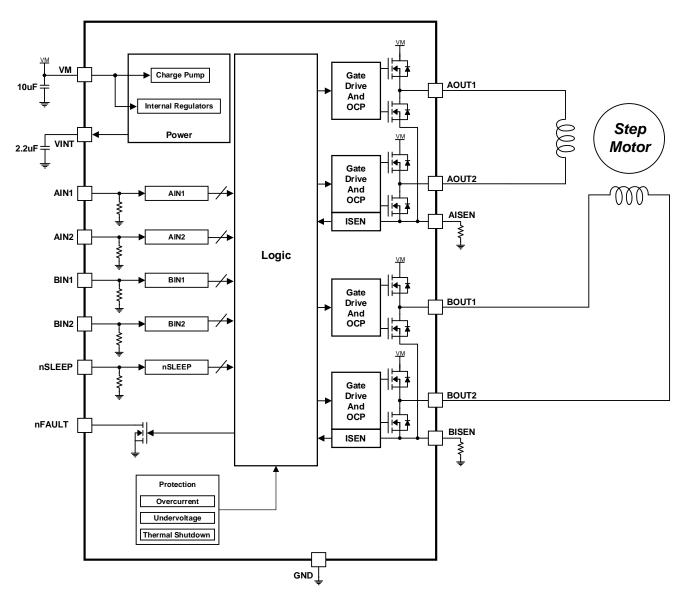


Figure 1. GD30DR3820 Functional Block Diagram



6.2 Operation

The GD30DR3820 device is an integrated motor driver solution for bipolar stepper motors. The device integrates two NMOS + NMOS H-bridges and current regulation circuitry. The GD30DR3820 can be powered with a supply voltage from 2.5V to 10.8V and can provide an output current up to 1.0A RMS.

A simple PWM interface allows easy interfacing to the controller circuit. The current regulation is a 23µs fixed offtime slow decay. The device includes a low-power sleep mode, which lets the system save power when not driving the motor.

6.2.1 PWM Motor Drivers

The GD30DR3820 contains drivers for two full H-bridges. The circuit block diagram is shown in Figure 2.

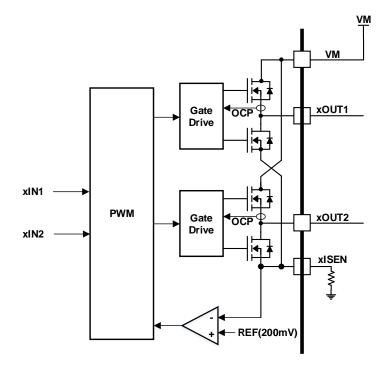


Figure 2. Motor Contorl Circuitry

6.2.2 Bridge Control and Decay Modes

Each H-Bridge has independent control pins and output pins, where AIN1 and AIN2 control the outputs of AOUT1 and AOUT2. Similarly, the BIN1 and BIN2 control the outputs of BOUT1 and BOUT2. The control logic is shown in Table 1.

xIN1	xIN2	xOUT1	xOUT2	FUNCTION
0	0	Z	Z	Coast/fast decay
0	1	L	Н	Reverse
1	0	Н	L	Forward
1	1	L	L	Brake/slow decay

Table 1. H-Bridge Control



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When the input signal is PWM, the speed of the motor can be justed. Since there is inductance in the motor winding, and the inductance has the characteristic that the current cannot mutate, when the drive current is interrupted, the current will find a way to continue flowing. The H-bridge provides two paths for current freewheeling. One is to disable the H-bridge and allow the current to flow through the body diode of the FET. This method is called fast decay mode. The other is to short-circuit the motor windings, which is called slow decay mode.

In summary, when one input pin is a PWM signal and the other pin is a low-level signal, it is fast decay mode. When the other pin is high-level signal, it is slow decay mode. See Table 2 for more information.

xIN1	xIN2	FUNCTION
PWM	0	Forward PWM, fast decay
1	PWM	Forward PWM, low decay
0	PWM	Reverse PWM, fast decay
PWM	1	Reverse PWM, low decay

Table 2. H-Bridge PWM Control

When disabling current control, the xISEN pin should be connected directly to ground.

Figure 3 shows the current paths in different drive and decay modes.

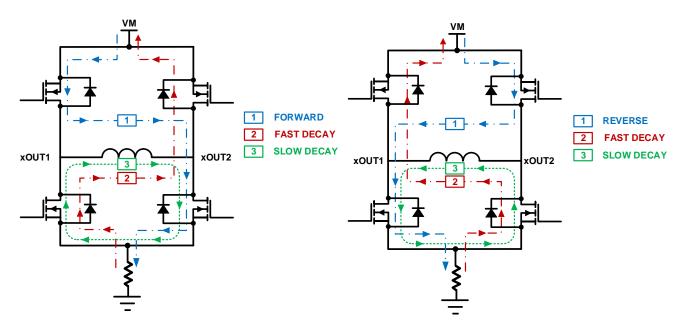


Figure 3. The driver outputs and current path

6.2.3 Current Control

The current on the motor winding can be adjusted by PWM with a certain frequency. This function can prevent excessive current from causing permanent damage to the device when the stepper motor is started or blocked.

The speed of the motor current climb depends on the voltage and winding inductance across the winding. When the device is operated with a motor, if the motor current reaches the set threshold, the H-bridge will be disabled before the next drive cycle starts. After the current is enabled, the voltage monitoring function on the xISEN pin is



enabled after a fixed blanking time.

The trip current threshold is set by a comparator that compares the voltage on the xISEN pin to an internal reference voltage(fixed at 200mv).

The tripping current is calculated as in Equation(1):

$$I_{\text{TRIP}} = \frac{200\text{mV}}{\text{R}_{\text{xISEN}}} \tag{1}$$

where: If the xISEN pin is connected to GND through a 400m Ω resistor, then the trip current will be 200mV / 400m Ω = 500mA.

NOTE: If current control is not needed, the xISEN pins should be connected directly to ground.

6.2.4 Decay Mode

When the motor current reaches the set threshold current, the H-bridge will enter the slow decay mode and reenable the H-bridge according to the status of the input signal after a fixed period of time(23us).

6.2.5 Slow Decay

In slow-decay mode, the high-side MOSFETs are turned off, while both of the low-side MOSFETs are turned on. The motor current flowing in the two-side MOSFETs will be decreased until the fixed off time is reached (typically 23 µs). Afterwards, in order to increase the winding current again, the MOSFETs on the high-side are enabled.

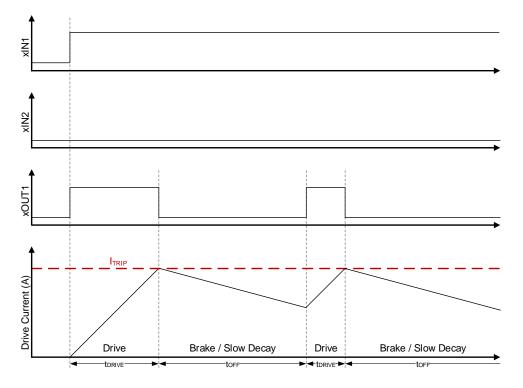


Figure 4. Slow Current Decay



6.2.6 Sleep Mode

Pulling down the nSLEEP puts the device into a low-power sleep state. In this state, the H-bridges are disabled, all internal logic is reset, all internal clocks stop working, and all inputs fail until nSLEEP is pulled up and activated. It takes some time, t_{WAKE}(typically 155us), from exit sleep mode to the motor driver becomes fully operational.

6.2.7 Protection Circuits

The GD30DR3820 is provided with overcurrent protection, overtemperature protection, and undervoltage protection functions.

6.2.7.1 VM undervoltage lockout(UVLO)

When the voltage of the VM pin is below the threshold voltage (V_{UVLO}), all circuits fail and all internal logic resets. The device returns to normal when the voltage of the VM pin rises above the threshold voltage (V_{UVLO}).

6.2.7.2 Thermal shutdown (TSD)

Once the die temperature exceeds the safe limits, all MOSFETs in the H-bridge will be disabled, and the nFAULT pin will also be driven low until the die temperature drops below the specified hysteresis (T_{HYS}). When the operation resumes, the nFAULT pin will be released.

6.2.7.3 Overcurrent protection (OCP)

Each FET has an analog current limit (I_{OCP}) circuit to limit the current through the FET, which will be realized by limiting the gate drive. All FETs in the H-bridge will be disabled and the nFAULT pin will be driven low if the analog current limiting duration exceeds the OCP deglitch time (t_{DEG}). After the OCP retry period (t_{OCP}), the driver will be re-enabled, and after the retry time, the nFAULT pin will be driven high. If the fault still exists, repeat the above operation. If the fault is eliminated, the normal operation will resum, and the nFAULT pin will remain deserted. Note that only the H-bridge that detect OCP will be disabled while other bridges work properly.

Short circuit to ground, supply, or across the motor winding will cause overcurrent shutdown, and overcurrent conditions are independently detected on both high-side and low-side devices. Please note that overcurrent protection does not use current detections circuit for PWM current control, so it can work even without xISEN resistors.

				-	
FAULT	CONDITION	ERROR REPORT	H-BRIDGE	INTERNAL CIRCUIT	RECOVERY
VM undervoltage (UVLO)	V _M < 2.4V	None	Disabled	Disabled	V _M > 2.5V
Overcurrent(OCP)	IOUT > IOCP	FAULTn	Disabled	Operating	OCP
Thermal Shutdown (TSD)	$T_J > T_{TSD}$	FAULTn	Disabled	Operating	$T_J < T_{TSD} - T_{HYS}$

6.3 Device Functional Modes

6.3.1 Device Enable

When the nSLEEP pin switches from high level to low level, after a period of time(t_{SLEEP} =10us), GD30DR3820 will enter sleep mode, and in this mode, the H-bridge will be in a disabled state(Hi-Z).



When the nSLEEP pin switches from low level to high level, The GD30DR3820 will automatically exit the sleep mode and enter the active mode after $t_{WAKE}(155us)$.

Table 4. Modes of Operation

OPERATING MODE	CONDITION	H-BRIDGE	INTERNAL CIRCUIT	
Operating	nSLEEP high Operating		Operating	
Sleep mode	nSLEEP low	Disabled	Disabled	
Fault encountered	Any fault condition met	Disabled	See Table 3	



7 Application Information

The GD30DR3820 is typically used to drive a bipolar stepper motor.

7.1 Typical Application Circuit

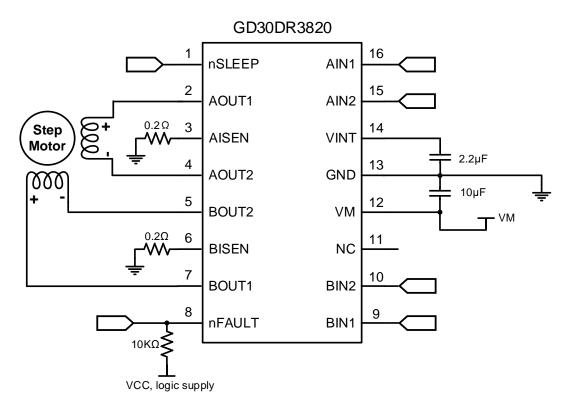


Figure 5. Schematic of GD30DR3820 Application

7.2 Design Example

For this design example, use the parameters in Table 5.

Table	5.	Design	Parameters
-------	----	--------	------------

PARAMETER	EXAMPLE VALUE		
Motor Supply Voltage	5V		
Logic Supply Voltage	3.3V		
Target RMS Current	0.5A		

7.3 Detailed Design Description

7.3.1 Motor Voltage

The motor voltage to use will depend on the ratings of the motor selected and the desired RPM. A higher voltage spins a bipolar stepper motor faster with the same PWM duty cycle applied to the power FETs. A higher voltage also increases the rate of current change through the inductive motor windings.



7.3.2 Motor Current Trip Point

When the voltage on pin xISEN exceeds VREF (0.2V), current regulation is activated. The RSENSE resistor connected from xISEN to ground, according to the Equation(1) to set the desired ITRIP level.

7.3.3 Sense Resistor

To ensure optimal performance, it is important for the sense resistor to be:

- Surface-mount
- Rated power meets the requirements
- Low Inductance
- Placed as close as to the motor driver pin

The power dissipated by the sense resistor can be calculated using $(I_{RMS})^2 \times R$. For example, if peak motor current is 0.8A, RMS motor current is 0.5A, and a 0.2 Ω sense resistor is used, the resistor dissipates 0.5A² × 0.2 Ω = 0.05W. The power rapidly increases with higher current levels.

Resistors generally have a rated power within a specific range of ambient temperature and a reduced power curve at high ambient temperatures. When the PCB is shared with other components generating heat, the system designer should increase the margin. It is always best to measure the actual sense resistor temperature in a final system.

Power resistors are larger and more expensive than standard resistors, and it is common to use multiple standard resistors in parallel between the sense node and ground to distributes the current and heat dissipation.

7.4 Power Dissipation

Power dissipation in the GD30DR3820 is dominated by the power dissipated in the output FET resistance, or $R_{DS(on)}$. The DC power dissipation of one H-bridge can be roughly estimated by Equation(2).

$$P_{D} \approx I_{RMS}^{2} \times \left(R_{HS_{DS}(ON)} + R_{LS_{DS}(ON)}\right)$$
(2)

where

- P_D is the device power dissipation
- RHS_DS(ON) is the resistance of the high-side FET
- RLS_DS(ON) is the resistance of the low-side FET
- I_{RMS} is the RMS or DC output current being supplied to the load

The maximum amount of power that can be dissipated in the device is dependent on ambient temperature and heatsinking.

 $R_{DS(ON)}$ increases with temperature, so as the device heats, the power dissipation increases. This must be taken into consideration when sizing the heatsink.



7.5 Typical Application Curves

 $T_A = 25^{\circ}C$, unless otherwise noted.

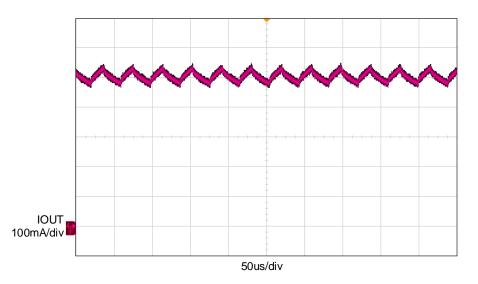


Figure 6. Current Regulation



8 Layout Guidelines and Example

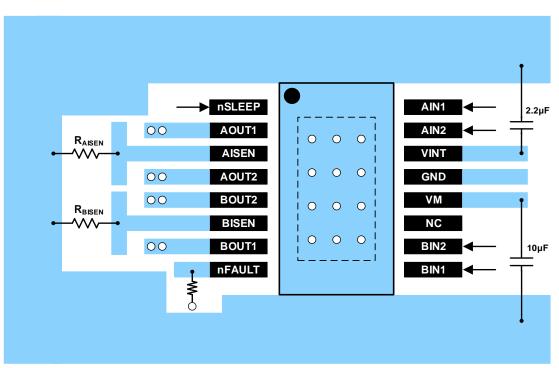
8.1 Layout Guidelines

Low ESR ceramic capacitors should be utilized for the VM to GND bypass capacitors. 10uF X5R or X7R types are recommended. These capacitors should be placed as close to the VM pins as possible with a thick trace or ground plane connection to the device GND pin.

In addition, bulk capacitor is required on the VM pin. This bulk capacitor can be ceramic or electrolytic type, but should also be placed as close as possible to the VM pin to minimize the loop inductance.

The high-current device outputs should use wide metal trace, and numerous vias should be used when connecting PCB layers.

Bypass VINT to ground with a 2.2uF ceramic capacitor rated 6.3V. Place this bypass capacitor as close to the pin as possible.



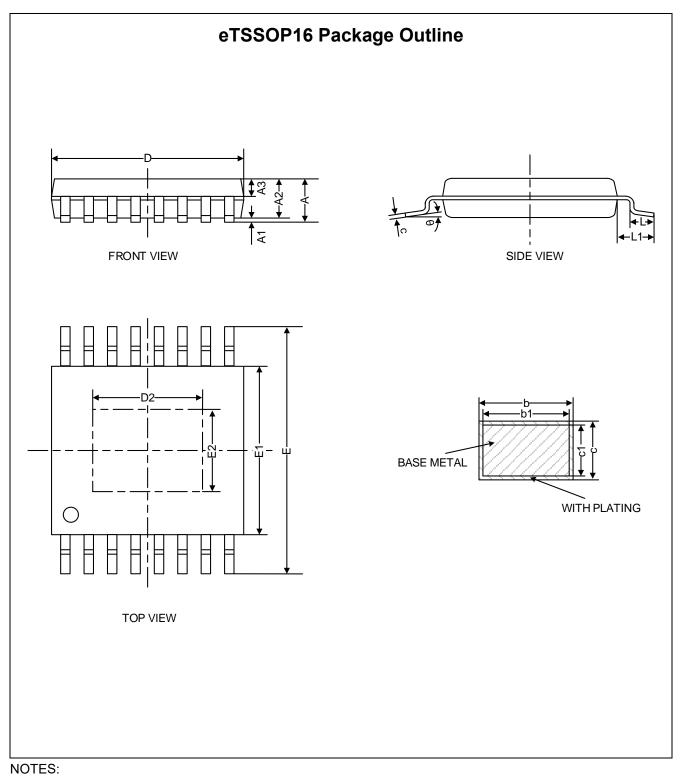
8.2 Layout Example

Figure 7. Typical Layout Example



9 Package Information

9.1 Outline Dimensions



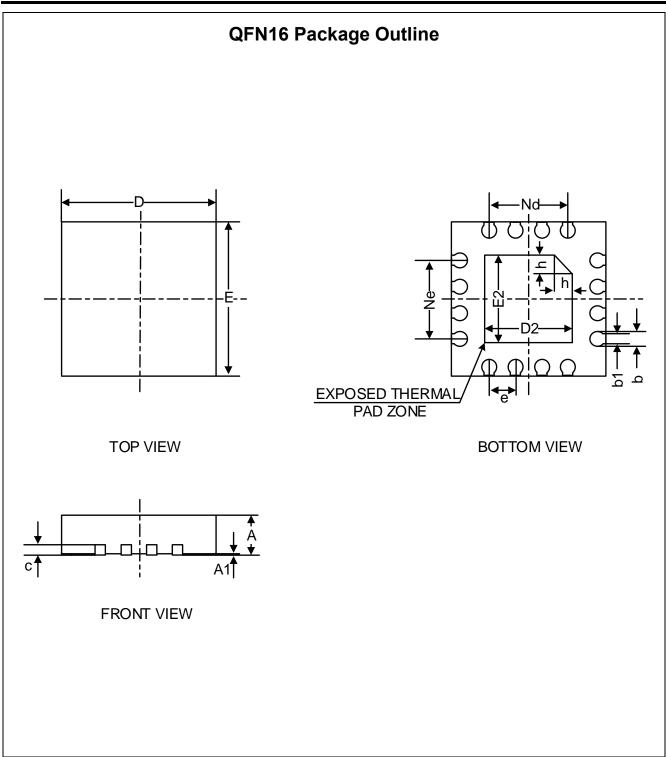
- 1. All dimensions are in millimeters.
- 2. Package dimensions does not include mold flash, protrusions, or gate burrs.
- 3. Refer to the Table 6 eTSSOP16 dimensions(mm).



Table 6. eTSSOP16 dimensions(mm)

SYMBOL	MIN	ТҮР	МАХ	
A			1.2	
A1	0.05		0.15	
A2	0.90	1.00	1.05	
A3	0.39	0.44	0.49	
b	0.20		0.28	
b1	0.19	0.22	0.25	
с	0.13		0.17	
c1	0.12	0.13	0.14	
D	4.90	5.00	5.10	
E	6.20	6.40	6.60	
E1	4.30	4.40	4.50	
e	0.65 BSC			
L	0.45		0.75	
L1	1.00 BSC			
θ	0°		8°	
D2	2.80 REF (L/F Size = 91 * 118)			
E2	2.10 REF (L/F Size = 91 * 118)			





NOTES: (continued)

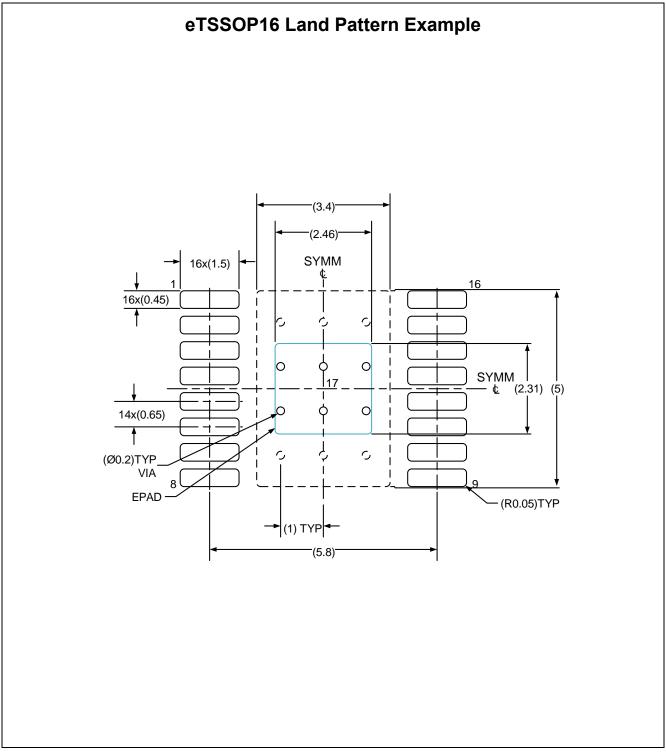
1. Refer to the Table 7 *QFN16 dimensions(mm)*.



SYMBOL	MIN	ТҮР	MAX	
A	0.70	0.75	0.80	
A1	0	0.02	0.05	
b	0.18	0.25	0.30	
b1	0.18 REF			
с	0.18 REF			
D	2.90	3.00	3.10	
D2	1.60	1.70	1.80	
е	0.50 BSC			
Ne	1.50 BSC			
Nd	1.50 BSC			
E	2.90	3.00	3.10	
E2	1.60	1.70	1.80	
L	0.25	0.30	0.35	
h	0.30	0.35	0.40	



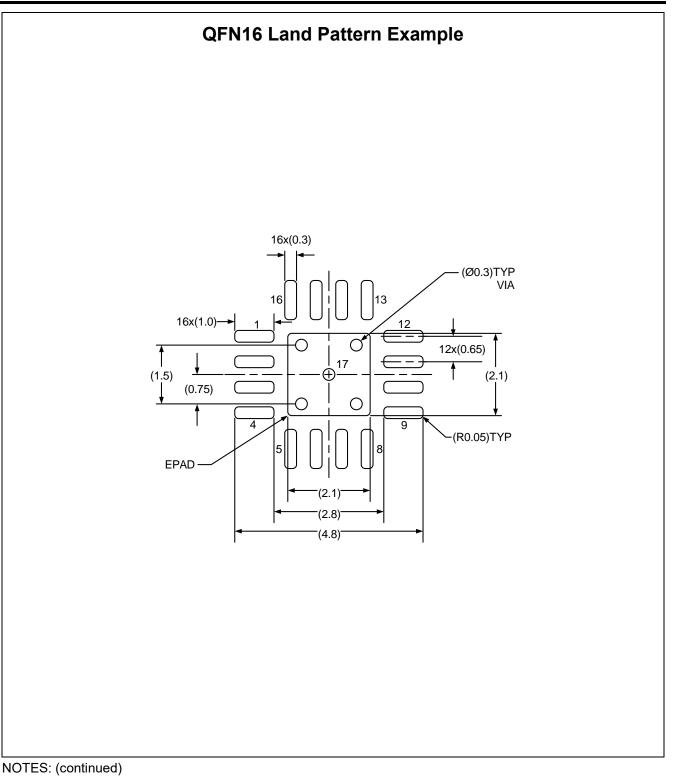
9.2 Recommended Land Pattern



NOTES: (continued)

- 1. Refer to the IPC-7351 can also help you complete the designs.
- 2. Explosed metal shown.
- 3. Drawing is 10X scale.





- 1. Refer to the IPC-7351 can also help you complete the designs.
- 2. Exposed metal shown.
- 3. Drawing is 10X scale.



10 Ordering Information

Ordering Code	Package Type	ECO Plan	Packing Type	MOQ	OP Temp(°C)
GD30DR3820LPTR-K	eTSSOP16	Green	Tape & Reel	3000	Industrial
					-40°C to +85°C
	QFN16	Green	Tape & Reel	3000	Industrial
GD30DR3820LUTR-K					-40°C to +85°C



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